

Generalization of scattering problem in a quasi 1D

We consider a device connected to the left and right leads as depicted below.

(Lead 1)



(Lead 2)

It is assumed that electrons flow steadily from the left to right. Then, the time-dependent Schrödinger equation can be reduced to the following form.

$$\begin{pmatrix} H_1 & \tau_1 & 0 \\ \tau_1^+ & H_d & \tau_2^+ \\ 0 & \tau_2 & H_2 \end{pmatrix} \begin{pmatrix} |\psi_i\rangle \\ |\psi_d\rangle \\ |\psi_2\rangle \end{pmatrix} = E \begin{pmatrix} |\psi_i\rangle \\ |\psi_d\rangle \\ |\psi_2\rangle \end{pmatrix} \quad \dots \quad (1)$$

$H, H_d, H_2, \tau_1, \tau_2$ are all matrices

At the beginning, we assume that the Schrödinger eq. of the isolated lead 1 is solved as

$$H_1 |\phi_{in}\rangle = E |\phi_{in}\rangle \quad \dots \quad (2)$$

Then, we consider ϕ_{in} as incident wave to the device region. In the steady state, ψ can be expressed by the sum of incident and reflected waves as

$$|\psi\rangle = \begin{pmatrix} |\psi_i\rangle \\ |\psi_d\rangle \\ |\psi_2\rangle \end{pmatrix} = |\phi_{in}\rangle + |x\rangle = \begin{pmatrix} |\phi_{in}\rangle + |x_i\rangle \\ |x_d\rangle \\ |x_2\rangle \end{pmatrix} \quad \dots \quad (3)$$

By inserting Eq. (3) into Eq. (1), we have

(2)

$$\begin{pmatrix} E - H_1 & -\tau_1 & 0 \\ -\tau_1^* & E - H_d & -\tau_2^* \\ 0 & -\tau_2 & E - H_2 \end{pmatrix} \begin{pmatrix} |\phi_{in}\rangle + |\chi_1\rangle \\ |\chi_d\rangle \\ |\chi_2\rangle \end{pmatrix} = \begin{pmatrix} 0 \\ 0 \\ 0 \end{pmatrix} \quad \dots (4)$$

→

$$\begin{pmatrix} E - H_1 & -\tau_1 & 0 \\ -\tau_1^* & E - H_d & -\tau_2^* \\ 0 & -\tau_2 & E - H_2 \end{pmatrix} \begin{pmatrix} |\chi_1\rangle \\ |\chi_d\rangle \\ |\chi_2\rangle \end{pmatrix} = \begin{pmatrix} -(E - H_1)|\phi_{in}\rangle \\ \tau_1^*|\phi_{in}\rangle \\ 0 \end{pmatrix} \quad \dots (5)$$

From Eq. (2), we see $-(E - H_1)|\phi_{in}\rangle = -(E - E)|\phi_{in}\rangle = 0$.

So, we have

$$\begin{pmatrix} |\chi_1\rangle \\ |\chi_d\rangle \\ |\chi_2\rangle \end{pmatrix} = G \begin{pmatrix} 0 \\ \tau_1^*|\phi_{in}\rangle \\ 0 \end{pmatrix} \quad \dots (6)$$

$$G = \begin{pmatrix} E - H_1 & -\tau_1 & 0 \\ -\tau_1^* & E - H_d & -\tau_2^* \\ 0 & -\tau_2 & E - H_2 \end{pmatrix}^{-1} = \begin{pmatrix} G_{11} & G_{1d} & G_{12} \\ G_{d1} & G_d & G_{d2} \\ G_{21} & G_{2d} & G_2 \end{pmatrix} \quad \dots (7)$$

From the second row in Eq. (6), we obtain

$$|\chi_d\rangle = G_d \tau_1^* |\phi_{in}\rangle = |\psi_d\rangle \quad \dots (8)$$

From the third row of Eq. (4), we have

(3)

$$(E - H_2) |x_2\rangle = \tau_2 |x_d\rangle$$

$g_2 = (E - H_2)^{-1}$
Note $g_2 \neq G_2$.

$$|x_2\rangle = (E - H_2)^{-1} \tau_2 |x_d\rangle = g_2 \tau_2 |x_d\rangle \quad \dots \dots (9)$$

Inserting Eq. (8) into Eq. (9), we obtain

$$|x_2\rangle = g_2 \tau_2 G_d \tau_1^+ |\phi_{in}\rangle = |\psi_2\rangle \quad \dots \dots (10)$$

From the first row of Eq. (5), we have

$$(E - H_1) |x_1\rangle = \tau_1 |x_d\rangle$$

$g_1 = (E - H_1)^{-1}$

$$|x_1\rangle = (E - H_1)^{-1} \tau_1 |x_d\rangle = g_1 \tau_1 |x_d\rangle \quad \dots \dots (11)$$

Inserting Eq. (8) into Eq. (11) yields the following relation.

$$|x_1\rangle = g_1 \tau_1 G_d \tau_1^+ |\phi_{in}\rangle \quad \dots \dots (12)$$

From Eqs. (8), (10), and (12), we obtain

$$\boxed{\begin{aligned} |\psi_1\rangle &= (1 + g_1 \tau_1 G_d \tau_1^+) |\phi_{in}\rangle \\ |\psi_d\rangle &= G_d \tau_1^+ |\phi_{in}\rangle \quad \dots \dots (13) \\ |\psi_2\rangle &= g_2 \tau_2 G_d \tau_1^+ |\phi_{in}\rangle \end{aligned}}$$

We see that ψ can be expressed by the incident wave ϕ_{in} .

If the scattering process occurs independently, we should consider the scattering process from the lead 2 to lead 1 equally.

Charge density in the device

From Eq. (13), we see that the wave function $\Psi_{d,n}^{(1)} (= \Psi_d)$ is contributed by the incident wave ϕ_{in} from the lead 1. So, the density operator of the device region can be defined by

$$\begin{aligned}
 \hat{\rho}_d^{(1)} &= \int_{-\infty}^{\infty} dE \sum_n f(E, \mu_i) \delta(E - E_n) |\Psi_{d,n}^{(1)}\rangle \langle \Psi_{d,n}^{(1)}| \\
 &= \int_{-\infty}^{\infty} dE f(E, \mu_i) \sum_n \delta(E - E_n) G_d \bar{\tau}_i^+ |\phi_{in}\rangle \langle \phi_{in}| \bar{\tau}_i G_d^+ \quad \xrightarrow{\text{Eq. (13)}} \frac{a_1}{2\pi} \\
 &= \int_{-\infty}^{\infty} dE f(E, \mu_i) G_d \bar{\tau}_i^+ \left[\sum_n \delta(E - E_n) |\phi_{in}\rangle \langle \phi_{in}| \right] \bar{\tau}_i G_d^+ \\
 &= \int_{-\infty}^{\infty} dE f(E, \mu_i) G_d \bar{\tau}_i^+ \frac{a_1}{2\pi} \bar{\tau}_i G_d^+ \quad \cdots \cdots \quad (14)
 \end{aligned}$$

By defining

$$\bar{\rho}_i = \bar{\tau}_i^+ a_1 \bar{\tau}_i \quad \cdots \cdots \quad (15)$$

We rewrite Eq. (14) as

$$\hat{\rho}_d^{(1)} = \frac{1}{2\pi} \int_{-\infty}^{\infty} dE f(E, \mu_i) G_d \bar{\rho}_i G_d^+ \quad \text{This is called Non-Equilibrium Green's Function (NEGF).} \quad \cdots \cdots \quad (16)$$

The matrix elements of $\hat{\rho}_d^{(1)}$ are evaluated by $\langle l | \hat{\rho}_d^{(1)} | m \rangle$ where $|l\rangle$ and $|m\rangle$ are basis functions in the device region.

By considering all the contributions from all the leads connected to the device, we obtain

Spin degeneracy

$$\hat{\rho} = \frac{2}{2\pi} \int_{-\infty}^{\infty} dE \left[\sum_i f(E, \mu_i) G_d \bar{\rho}_i G_d^+ \right]$$

\$\rightarrow\$

$\cdots \cdots \quad (17)$

Current through the device

In the non-equilibrium steady state the probability density of the device region is conserved. We evaluate the flux of the probability density using the time-dependent Schrödinger equation as

$$i\hbar \frac{\partial}{\partial t} \begin{pmatrix} |\Psi_i\rangle \\ |\Psi_d\rangle \\ |\Psi_e\rangle \end{pmatrix} = \begin{pmatrix} H_i & \tau_i & 0 \\ \tau_i^+ & H_d & \tau_2^+ \\ 0 & \tau_2 & H_e \end{pmatrix} \begin{pmatrix} |\Psi_i\rangle \\ |\Psi_d\rangle \\ |\Psi_e\rangle \end{pmatrix} \quad \dots \dots (18)$$

The time evolution of the integrated probability density is given by

$$\begin{aligned} \frac{\partial}{\partial t} \langle \Psi_d | \Psi_d \rangle &= \frac{\partial \langle \Psi_d |}{\partial t} |\Psi_d\rangle + \langle \Psi_d | \frac{\partial |\Psi_d\rangle}{\partial t} \\ &= \frac{i}{\hbar} \left[\langle \Psi_d | \tau_2 + \langle \Psi_d | H_d + \langle \Psi_i | \tau_1 \right] |\Psi_d\rangle \quad \leftarrow \text{From Eq. (18)} \\ &\quad - \frac{i}{\hbar} \langle \Psi_d | \left[\tau_1^+ |\Psi_i\rangle + H_d |\Psi_d\rangle + \tau_2^+ |\Psi_e\rangle \right] \\ &= \frac{i}{\hbar} \left[(\langle \Psi_i | \tau_1 | \Psi_d \rangle - \langle \Psi_d | \tau_1^+ | \Psi_i \rangle) + (\langle \Psi_2 | \tau_2 | \Psi_d \rangle - \langle \Psi_d | \tau_2^+ | \Psi_2 \rangle) \right] \end{aligned} \quad \dots \dots (19)$$

Each term in Eq. (19) can be regarded as

the contribution from each lead k .

The flow of

$$\text{current is } i_k = -\frac{ie}{\hbar} (\underbrace{\langle \Psi_k | \tau_k | \Psi_d \rangle - \langle \Psi_d | \tau_k^+ | \Psi_k \rangle}_{\text{Because of conservation of probability density,}}) \quad \text{This is purely imaginary.} \quad \dots \dots (20)$$

opposite to that of electron

we have

$$\sum_k i_k = 0 \quad \dots \dots (21)$$

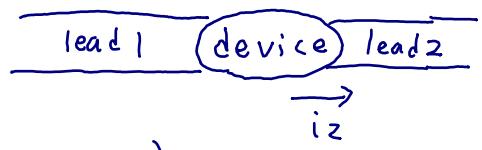
(6)

From the condition of Eq. (21) for the steady state,
we obtain

$$i_2 = -i_1 \quad \dots \dots \quad (22)$$

So, the current from the lead 1 to the lead 2

can be calculated from i_2 .



$$i_{1 \rightarrow 2} = i_2 = -\frac{ie}{\hbar} \left(\langle \psi_2 | \tau_2 | \psi_d \rangle - \langle \psi_d | \tau_2^\dagger | \psi_2 \rangle \right) \quad \dots \dots \quad (23)$$

Using Eq. (13), one can calculate as

$$\begin{aligned} i_{1 \rightarrow 2} &= -\frac{ie}{\hbar} \left(\langle \phi_{in} | \tau_1 G_d^\dagger \tau_2^\dagger g_2^\dagger \tau_2 G_d \tau_1^\dagger | \phi_{in} \rangle - \langle \phi_{in} | \tau_1 G_d^\dagger \tau_2^\dagger g_2 \tau_2 G_d \tau_1^\dagger | \phi_{in} \rangle \right) \\ &= -\frac{ie}{\hbar} \langle \phi_{in} | \tau_1 G_d^\dagger \tau_2^\dagger (g_2^\dagger - g_2) \tau_2 G_d \tau_1^\dagger | \phi_{in} \rangle \quad \dots \dots \quad (24) \end{aligned}$$

By defining

$$\Gamma_2 = -i \tau_2^\dagger (g_2^\dagger - g_2) \tau_2 \quad \dots \dots \quad (25)$$

Eq. (24) can be rewritten as

$$i_{1 \rightarrow 2} = \frac{e}{\hbar} \langle \phi_{in} | \tau_1 G_d^\dagger \Gamma_2 G_d \tau_1^\dagger | \phi_{in} \rangle \quad \dots \dots \quad (26)$$

* Equivalence of Eqs. (15) and (25).

$$\text{Noting } g_k(E) = \sum_n \frac{|\phi_{kn}\rangle \langle \phi_{kn}|}{E - E_n} \quad \dots \dots \quad (27)$$

$$g_k(E) - g_k^\dagger(E) = \sum_n |\phi_{kn}\rangle \langle \phi_{kn}| \left(\frac{1}{E - E_n + i\eta} - \frac{1}{E - E_n - i\eta} \right)$$

$$= \sum_n |\phi_{kn}\rangle \langle \phi_{kn}| \left(\frac{-2i\eta}{(E - E_n)^2 + \eta^2} \right) \xrightarrow{\eta \rightarrow 0} \sum_n |\phi_{kn}\rangle \langle \phi_{kn}| (-2i\pi \delta(E - E_n))$$

So, one can calculate as

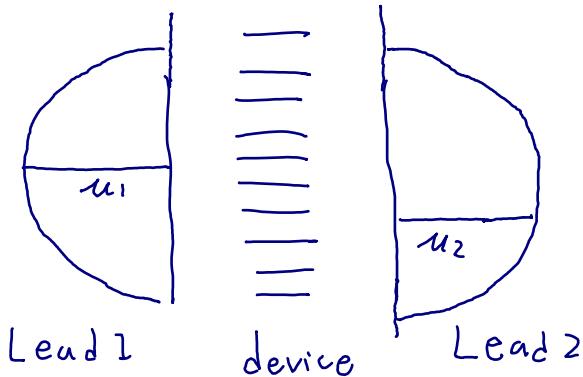
$$\Gamma_k = -i \tau_k^\dagger (g_k^\dagger - g_k) \tau_k = \tau_k^\dagger \left(\sum_n 2\pi \delta(E - E_n) |\phi_{kn}\rangle \langle \phi_{kn}| \right) \tau_k$$

Thus, we see the definition of Eq. (25) is equivalent to Eq. (15).

Now we consider all the contributions from the incident wave ϕ_{in} to the current. Eq. 26)

$$\begin{aligned}
 I_{1 \rightarrow 2} &= 2 \frac{e}{\hbar} \int_{-\infty}^{\infty} dE f(E, u_1) \sum_n \delta(E - E_n) \langle \phi_{in} | \tau_1 G_d^\dagger \Gamma_2 G_d \tau_1^\dagger | \phi_{in} \rangle \\
 &\quad \text{Spin multiplicity} \qquad \qquad \qquad \text{Identity operator in the device region} \\
 &= \frac{ze}{\hbar} \int_{-\infty}^{\infty} dE f(E, u_1) \sum_n \sum_m \delta(E - E_n) \langle \phi_{in} | \tau_1 | m \rangle \langle m | G_d^\dagger \Gamma_2 G_d \tau_1^\dagger | \phi_{in} \rangle \\
 &= \frac{ze}{\hbar} \int_{-\infty}^{\infty} dE f(E, u_1) \sum_m \langle m | G_d^\dagger \Gamma_2 G_d \tau_1^\dagger \left(\sum_n \delta(E - E_n) |\phi_{in}\rangle \langle \phi_{in}| \right) \tau_1 | m \rangle \\
 &= \frac{ze}{\hbar} \int_{-\infty}^{\infty} dE f(E, u_1) \sum_m \langle m | G_d^\dagger \Gamma_2 G_d \tau_1^\dagger \frac{a_1}{2\pi} \tau_1 | m \rangle \quad \text{See Eq. (14)} \\
 &= \frac{e}{\pi \hbar} \int_{-\infty}^{\infty} dE f(E, u_1) \text{Tr} (G_d^\dagger \Gamma_2 G_d \tau_1^\dagger) \quad \dots \dots \quad (28)
 \end{aligned}$$

By considering the following situation:



Transmission
 $T(E) = \text{Tr} (G_d^\dagger \Gamma_2 G_d \tau_1^\dagger) \quad \dots \dots \quad (30)$

The expressions of Eqs. (29) and (30) are called Landauer formula.

it is found that $I_{1 \rightarrow 2}$ is regarded the current from the lead 1 to the lead 2. As well we need to take account of the current from the lead 2 to lead 1.

Since we assume the steady state of Eq. (22), we have the total current as

$$I = \frac{2e}{\hbar} \int_{-\infty}^{\infty} dE (f(E, u_1) - f(E, u_2)) \text{Tr} (G_d^\dagger \Gamma_2 G_d \tau_1^\dagger) \quad \dots \dots \quad (29)$$

Evaluation of G_d

The Green's function G_d of the device region is evaluated by Eq.(7). By definition, we have

$$\begin{pmatrix} E - H_1 & -\tau_1 & 0 \\ -\tau_1^+ & E - H_d & -\tau_2^+ \\ 0 & -\tau_2 & E - H_2 \end{pmatrix} \begin{pmatrix} G_1 & G_{1d} & G_{1z} \\ G_{d1} & G_d & G_{d2} \\ G_{z1} & G_{zd} & G_z \end{pmatrix} = \begin{pmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{pmatrix} \quad \dots (31)$$

row 1 x column 2

$$(E - H_1) G_{1d} - \tau_1 G_d = 0 \quad \dots (32)$$

row 2 x column 2

$$-\tau_1^+ G_{1d} + (E - H_d) G_d - \tau_2^+ G_{zd} = 1 \quad \dots (33)$$

row 3 x column 2

$$-\tau_2 G_d + (E - H_2) G_{zd} = 0 \quad \dots (34)$$

From (32), we have

$$G_{1d} = (E - H_1)^{-1} \tau_1 G_d = g_1 \tau_1 G_d \quad \dots (35)$$

From (34), we have

$$G_{zd} = (E - H_2)^{-1} \tau_2 G_d = g_2 \tau_2 G_d \quad \dots (36)$$

Inserting Eqs. (35) and (36) into Eq. (33), we obtain

$$-\tau_1^+ g_1 \tau_1 G_d + (E - H_d) G_d - \tau_2^+ g_2 \tau_2 G_d = 1$$

$$(E - H_d - \Sigma_1 - \Sigma_2) G_d = 1$$

$$G_d = (E - H_d - \Sigma_1 - \Sigma_2)^{-1} \quad \dots (37)$$

where

$$\Sigma_1 = \tau_1^+ g_1 \tau_1 \quad \dots (38)$$

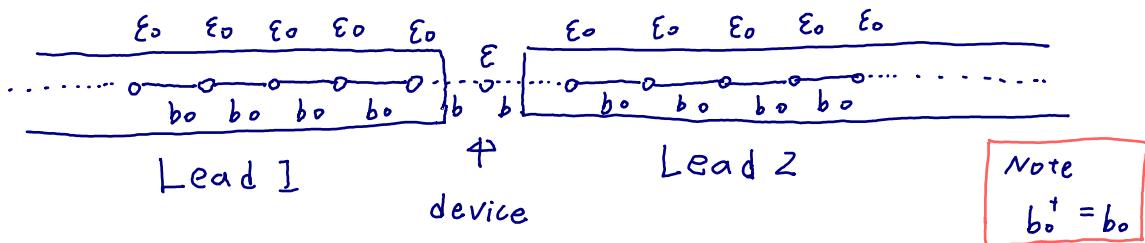
$$\Sigma_2 = \tau_2^+ g_2 \tau_2 \quad \dots (39)$$

g_1 and g_2 are called surface Green's function.

Σ_1 and Σ_2 are called self-energy.

An example: linear chain model

Let's consider a linear chain model as shown below.



To calculate the surface Green functions of the leads 1 and 2, we employ the recursion method discussed in "Recursion_method.pdf".

From the page 13 in the lecture notes, we have

$$g_1(z) = g_2(z) = \frac{z - \varepsilon_0 - \sqrt{(z - \varepsilon_0)^2 - 4b_0^2}}{2b_0^2} \quad \dots \quad (40)$$

The self-energies of Eqs. (38) and (39) are given by

$$\Sigma_1(z) = \Sigma_2(z) = b^2 g_1(z) = \frac{b^2}{2b_0^2} (z - \varepsilon_0 - \sqrt{(z - \varepsilon_0)^2 - 4b_0^2}) \quad \dots \quad (41)$$

The Green function of Eq. (37) is found to be

$$G_d(z) = \left[z - \varepsilon - \frac{b^2}{b_0^2} (z - \varepsilon_0 - \sqrt{(z - \varepsilon_0)^2 - 4b_0^2}) \right]^{-1} \quad \dots \quad (42)$$

Also, Γ_1 and Γ_2 are calculated as

$$\Gamma_1 = \Gamma_2 = i b (g_1 - g_1^+) b = \frac{i b^2}{2b_0^2} (z - \cancel{\varepsilon_0} - \sqrt{(z - \varepsilon_0)^2 - 4b_0^2} - z^* + \cancel{\varepsilon_0} + \sqrt{(z^* - \varepsilon_0)^2 - 4b_0^2})$$

Thus, the transmission is given by

$$T(E) = \text{Tr} (G_d^+ \Gamma_2 G_d \Gamma_1) \quad \dots \quad (44)$$

$\uparrow \quad \uparrow \quad \uparrow \quad \uparrow$
 (42) (43) (42) (43)

In case that $\varepsilon = \varepsilon_0$ and $h = h_0$, we have

(10)

$$G_d(z) = \frac{1}{\sqrt{(z - \varepsilon_0)^2 - 4b_0^2}} \quad \dots \dots (45)$$

This is equivalent to Eq. (28) in the page 16 of lecture notes 2.

Taking $z = E + i\eta$ ($\eta \rightarrow 0$), G_d becomes

$$G_d(E + i\eta) = \frac{1}{\sqrt{(E - \varepsilon_0)^2 - \eta^2 + 2i\eta(E - \varepsilon_0) - 4b_0^2}}$$

For $(E - \varepsilon_0)^2 - 4b_0^2 < 0 \Rightarrow \varepsilon_0 - 2|b_0| < E < \varepsilon_0 + 2|b_0|$

$$\text{Im } G_d = -\frac{1}{\sqrt{4b_0^2 - (E - \varepsilon_0)^2}}, \quad \text{Re } G_d = 0 \quad \dots \dots (46)$$

For $E < \varepsilon_0 - 2|b_0|$, $\varepsilon_0 + 2|b_0| < E$

$$\text{Im } G_d = 0, \quad \text{Re } G_d = \frac{1}{\sqrt{(E - \varepsilon_0)^2 - 4b_0^2}} \quad \dots \dots (47)$$

Noting for $\eta \rightarrow 0$

$$\begin{aligned} P_1 &= P_2 \\ &= i b_0^2 (g_1 - g_1^+) = -2 b_0^2 i_m g_1 = \begin{cases} \sqrt{4b_0^2 - (E - \varepsilon_0)^2} & \varepsilon_0 - 2|b_0| < E < \varepsilon_0 + 2|b_0| \\ 0 & \text{else} \end{cases} \end{aligned} \quad \dots \dots (48)$$

From Eqs. (46) - (48), we obtain the transmission.

$$T(E) = \overline{T}_r (G_d^+ P_2 G_d P_1)$$

$$= (-i) \times i \left(\frac{1}{\sqrt{4b_0^2 - (E - \varepsilon_0)^2}} \right)^2 \times \left(\sqrt{4b_0^2 - (E - \varepsilon_0)^2} \right)^2 = 1$$

for $\varepsilon_0 - 2|b_0| < E < \varepsilon_0 + 2|b_0|$

Otherwise

$\dots \dots (49)$

$$T(E) = 0$$

Conductance quantum

Let us start Eq. (29).

$$I = \frac{2e}{h} \int_{-\infty}^{\infty} dE T(E) [f(E-\mu_1) - f(E-\mu_2)] \quad \dots \dots (29)$$

Our purpose here is to find a relation

$$I = \frac{V}{R} = VG \quad \dots \dots (50)$$

where V is the source-drain bias voltage, R resistance, and $G \equiv \frac{1}{R}$ conductance. To derive Eq. (50) based on Eq. (29), we consider a case that $\mu_1 = \mu_2 + eV = \mu + eV$ with a tiny V .

Then, Eq. (29) can be approximated by

$$\begin{aligned} I &\approx \frac{2e}{h} \int_{-\infty}^{\infty} dE \left[T(\mu) + \left. \frac{\partial T(E)}{\partial E} \right|_{E=\mu} (E-\mu) \right] \left[f(E-\mu) + \left. \frac{\partial f(E-(\mu+eV))}{\partial V} \right|_{V=0} V - f(E-\mu) \right] \\ &= \frac{2e}{h} \left[\int_{-\infty}^{\infty} dE T(\mu) \frac{\partial f(E-(\mu+eV))}{\partial V} V + \int_{-\infty}^{\infty} dE \frac{\partial T(E)}{\partial E} (E-\mu) \frac{\partial f(E-(\mu+eV))}{\partial V} V \right] \end{aligned}$$

$$\text{Noting } \frac{\partial f(E-(\mu+eV))}{\partial V} = -e \frac{\partial f(E-(\mu+eV))}{\partial E} \xrightarrow{V \rightarrow 0} e \delta(E-\mu) \quad \dots \dots (51)$$

Eq. (51) becomes

$$I = \frac{2e^2}{h} T(\mu) V \int_{-\infty}^{\infty} dE \delta(E-\mu) = V \left(T(\mu) \frac{2e^2}{h} \right) \quad \dots \dots (52)$$

Compared to Eq. (50), we obtain

G_0 is called

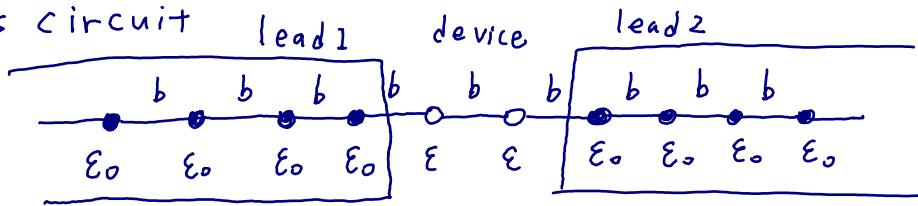
$$G = T(\mu) G_0 \quad \dots \dots (53)$$

conductance quantum

$$G_0 = \frac{2e^2}{h} \quad \dots \dots (54)$$

Conductance of series and parallel circuits

(a) Series circuit



$$\epsilon_0 = 0$$

As well as the lecture notes 2 and 13, the surface Green functions are given by

$$g_1(z) = g_2(z) = \frac{z - \sqrt{z^2 - 4b^2}}{2b^2} \quad \dots \dots (1)$$

The self-energies can be calculated as

$$\Sigma_1(z) = \begin{pmatrix} b \\ 0 \end{pmatrix} g_1(z) (b, 0) = \begin{pmatrix} b^2 g_1(z) & 0 \\ 0 & 0 \end{pmatrix} \quad \dots \dots (2)$$

$$\Sigma_2(z) = \begin{pmatrix} 0 \\ b \end{pmatrix} g_2(z) (0, b) = \begin{pmatrix} 0 & 0 \\ 0 & b^2 g_2(z) \end{pmatrix} \quad \dots \dots (3)$$

The Green function of the device region is given by

$$G_d(z) = (z - H_d - \Sigma_1(z) - \Sigma_2(z))^{-1}$$

$$= \begin{pmatrix} z - \varepsilon - \frac{1}{z}(z - \sqrt{z^2 - 4b^2}) & -b \\ -b & z - \varepsilon - \frac{1}{z}(z - \sqrt{z^2 - 4b^2}) \end{pmatrix}^{-1} = \begin{pmatrix} A & B \\ B & A \end{pmatrix}^{-1} = \frac{1}{A^2 - B^2} \begin{pmatrix} A & -B \\ -B & A \end{pmatrix}$$

$$= \frac{1}{C} \begin{pmatrix} z - \varepsilon - \frac{1}{z}(z - \sqrt{z^2 - 4b^2}) & b \\ b & z - \varepsilon - \frac{1}{z}(z - \sqrt{b^2 - 4b^2}) \end{pmatrix} \quad C = [z - \varepsilon - \frac{1}{z}(z - \sqrt{z^2 - 4b^2})]^2 - b^2 \quad \dots \dots (5)$$

$$= \begin{pmatrix} \alpha & \beta \\ \beta & \alpha \end{pmatrix} \quad \dots \dots (4)$$

(13)

P_1 and P_2 can be calculated as

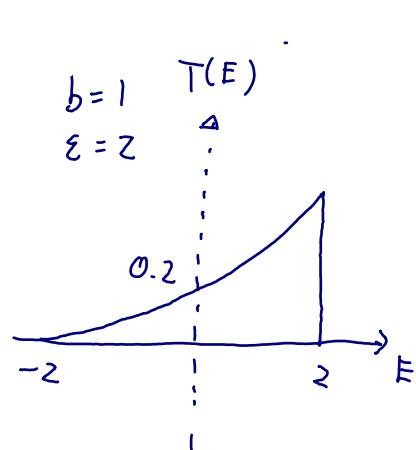
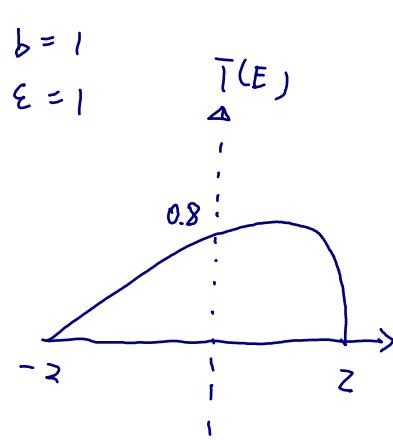
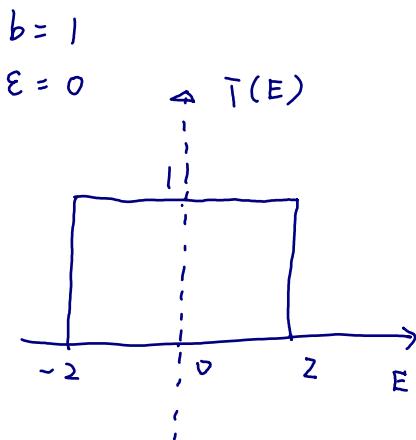
$$\begin{aligned}
 P_1 &= i \tau_1^+ (g_1 - g_1^+) \tau_1 = i \begin{pmatrix} b \\ 0 \end{pmatrix} 2i \operatorname{Im} g_1 (b \ 0) \\
 &= -2b^2 \operatorname{Im} g_1 \begin{pmatrix} 1 & 0 \\ 0 & 0 \end{pmatrix} = \sqrt{4b^2 - E^2} \begin{pmatrix} 1 & 0 \\ 0 & 0 \end{pmatrix} \\
 P_2 &= i \tau_2^+ (g_2 - g_2^+) \tau_1 = i \begin{pmatrix} 0 \\ b \end{pmatrix} 2i \operatorname{Im} g_2 (0 \ b) \\
 &= -2b^2 \operatorname{Im} g_2 \begin{pmatrix} 0 & 0 \\ 0 & 1 \end{pmatrix} = \sqrt{4b^2 - E^2} \begin{pmatrix} 0 & 0 \\ 0 & 1 \end{pmatrix}
 \end{aligned}$$

$$\text{for } -z|b| < E < z|b| \quad \dots \dots (6)$$

Otherwise, they are zero.

Then, one can calculate the transmission as

$$\begin{aligned}
 T(E) &= \operatorname{Tr}(G_d^\dagger P_2 G_d P_1) \\
 &= (4b^2 - E^2) \operatorname{Tr} \left[\begin{pmatrix} \alpha^* & \beta^* \\ \beta^* & \alpha^* \end{pmatrix} \begin{pmatrix} 0 & 0 \\ 0 & 1 \end{pmatrix} \begin{pmatrix} \alpha & \beta \\ \beta & \alpha \end{pmatrix} \begin{pmatrix} 1 & 0 \\ 0 & 0 \end{pmatrix} \right] \\
 &= (4b^2 - E^2) \operatorname{Tr} \left[\begin{pmatrix} 0 & \beta^* \\ 0 & \alpha^* \end{pmatrix} \begin{pmatrix} \alpha & 0 \\ \beta & 0 \end{pmatrix} \right] = (4b^2 - E^2) \operatorname{Tr} \begin{pmatrix} \beta^* \beta & 0 \\ \alpha^* \beta & 0 \end{pmatrix} \\
 &= (4b^2 - E^2) \beta^* \beta \\
 &= (4b^2 - E^2) \times b^2 \times \frac{1}{c^* c} \quad \dots \dots (7)
 \end{aligned}$$



Noting that $T(E)$ at $E=0$ is the conductance,

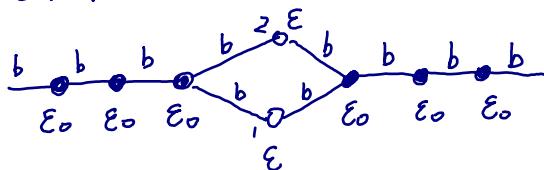
we have C^*C at $E=0$ as

$$C^*C(E=0) = [(-\varepsilon + ib)^2 - b^2] [(-\varepsilon - ib)^2 - b^2] \\ = 4b^4 + \varepsilon^4 \quad \dots \dots (8)$$

Thus, we obtain from Eqs. (7) and (8)

$$G = 4b^2 \times b^2 \times \frac{1}{C^*C(E=0)} \\ = \frac{4b^4}{4b^4 + \varepsilon^4} \quad \text{in } G_0 \quad \dots \dots (9)$$

(b) Parallel circuit



The surface Green functions are given by Eqs. (1).

The self-energies can be calculated as

$$\Sigma_1(z) = \begin{pmatrix} b \\ b \end{pmatrix} g_1(z) \begin{pmatrix} b & b \end{pmatrix} = b^2 g_1(z) \begin{pmatrix} 1 & 1 \\ 1 & 1 \end{pmatrix} \quad \dots \dots (10)$$

$$\Sigma_2(z) = \begin{pmatrix} b \\ b \end{pmatrix} g_2(z) \begin{pmatrix} b & b \end{pmatrix} = b^2 g_2(z) \begin{pmatrix} 1 & 1 \\ 1 & 1 \end{pmatrix} \quad \dots \dots (11)$$

The Green function of the device region is given by

$$G_d(z) = (zI - H_d - \Sigma_1 - \Sigma_2)^{-1}$$

$$= \begin{pmatrix} \cancel{z} - \varepsilon - \cancel{z} + \sqrt{z^2 - 4b^2} & -z + \sqrt{z^2 - 4b^2} \\ -z + \sqrt{z^2 - 4b^2} & \cancel{z} - \varepsilon - \cancel{z} + \sqrt{z^2 - 4b^2} \end{pmatrix}^{-1} = \begin{pmatrix} A & B \\ B & A \end{pmatrix}^{-1} = \frac{1}{A^2 - B^2} \begin{pmatrix} A & -B \\ -B & A \end{pmatrix} \quad \dots \dots (12)$$

$$G_d(z) = \frac{1}{c} \begin{pmatrix} -\varepsilon + \sqrt{z^2 - 4b^2} & z - \sqrt{z^2 - 4b^2} \\ z - \sqrt{z^2 - 4b^2} & -\varepsilon + \sqrt{z^2 - 4b^2} \end{pmatrix} \quad \dots \quad (13)$$

$$C = (-\varepsilon + \sqrt{z^2 - 4b^2})^2 - (z - \sqrt{z^2 - 4b^2})^2 \quad \dots \quad (14)$$

Γ_1 and Γ_2 can be calculated as

$$\begin{aligned} \Gamma_1 &= i \tau_1^+ (g_1 - g_1^+) \tau_1 \\ &= i \begin{pmatrix} b \\ b \end{pmatrix} z i \text{Im } g_1 \begin{pmatrix} b & b \end{pmatrix} = -2b^2 \text{Im } g_1 \begin{pmatrix} 1 & 1 \\ 1 & 1 \end{pmatrix} \\ &= \sqrt{4b^2 - E^2} \begin{pmatrix} 1 & 1 \\ 1 & 1 \end{pmatrix} \quad \dots \quad (15) \end{aligned}$$

$$\begin{aligned} \Gamma_2 &= i \bar{\tau}_2^+ (g_2 - g_2^+) \bar{\tau}_2 = -2b^2 \text{Im } g_2 \begin{pmatrix} 1 & 1 \\ 1 & 1 \end{pmatrix} \\ &= \sqrt{4b^2 - E^2} \begin{pmatrix} 1 & 1 \\ 1 & 1 \end{pmatrix} \quad \dots \quad (16) \end{aligned}$$

for $-2|b| < E < 2|b|$

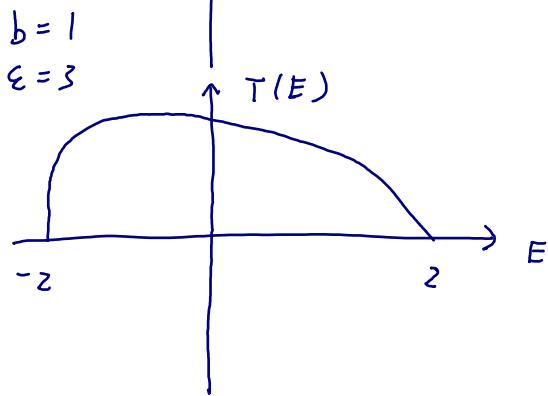
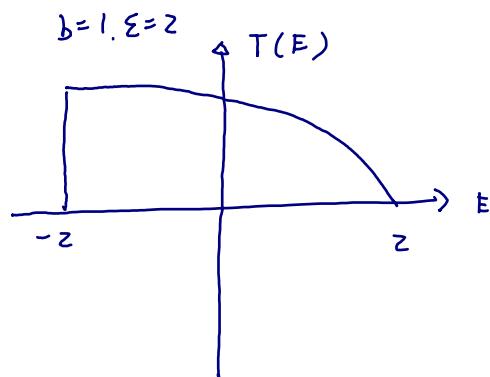
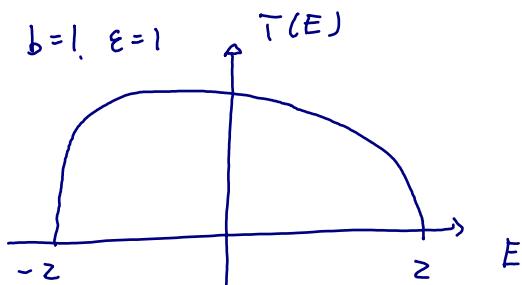
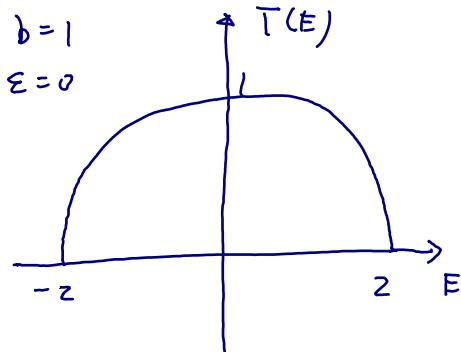
otherwise, they are zero.

So, the transmission are given by

$$\begin{aligned} T(E) &= \text{Tr} (G_d^\dagger \Gamma_2 G_d \Gamma_1) \\ &= \frac{1}{c^* c} (4b^2 - E^2) \text{Tr} \left[\begin{pmatrix} -\varepsilon + \sqrt{z^2 - 4b^2} & z^* - \sqrt{z^2 - 4b^2} \\ z^* - \sqrt{z^2 - 4b^2} & -\varepsilon + \sqrt{z^2 - 4b^2} \end{pmatrix} \begin{pmatrix} 1 & 1 \\ 1 & 1 \end{pmatrix} \right. \\ &\quad \times \left. \begin{pmatrix} -\varepsilon + \sqrt{z^2 - 4b^2} & z - \sqrt{z^2 - 4b^2} \\ z - \sqrt{z^2 - 4b^2} & -\varepsilon + \sqrt{z^2 - 4b^2} \end{pmatrix} \begin{pmatrix} 1 & 1 \\ 1 & 1 \end{pmatrix} \right] \end{aligned}$$

(16)

$$T(E) = \frac{-4E^2 + 16b^2}{(\varepsilon + z - 2\sqrt{z^2 - 4b^2})(\varepsilon + z^* - 2\sqrt{z^{*2} - 4b^2})} \quad \dots (17)$$



From Eq.(17) at $E=0$, we obtain the conductance

$$G = \frac{16b^2}{(\varepsilon - z \times zbi)(\varepsilon + z \times zbi)} = \frac{16b^2}{\varepsilon^2 + 16b^2} \quad \dots (18)$$

in G_0 .

Conductance of



$$\frac{4b^2}{(-\varepsilon + z b)^2} \quad \dots (19)$$

Comparing with Eqs. (9), (18), and (19), we see that

Ohm's law is not valid anymore.